

# Lightning Characteristics Analysis of Grounding Devices by Modified Partial Element Equivalent Circuit Method

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**Abstract**—This paper proposes a time-domain method for the lightning transient performance of the grounding device. Based on the partial element equivalent circuit method, it considers the frequency-dependence in the time-domain, and refers to the alternating direction implicit difference scheme for the unconditionally stable solution.

**Keywords**- grounding; partial element equivalent circuit (PEEC); time-domain; frequency-dependent; lightning.

## I. INTRODUCTION

The lightning transient characteristic of the grounding device plays an important role in the lightning protection [1]. For evaluating and analyzing, simulation is a very effective approach. There are several kinds of methods such as the transmission line method, the finite-different time-domain method, the method of moment and the PEEC method. Compared with the former three, PEEC gives consideration to both efficiency and accuracy [2].

In this paper, a modified PEEC method is proposed with following characteristic: a) bases on the electromagnetic quasi-static (EMQS) hypothesis and time-domain analysis; b) considers the frequency-dependence, as well as the multi-layer soil and the mutual coupling; c) refers the ADI difference scheme for the unconditionally stable solution.

## II. GENERAL PRINCIPLE OF METHOD

Firstly, for conductors buried in the lossy ground, the node voltages  $\mathbf{V}$  and the branch currents  $\mathbf{I}$  are arranged alternatively,  $\mathbf{V}$  are located at the nodes, and  $\mathbf{I}$  are located at the middle of the branches. Then, the EMQS equivalent circuit is established in the frequency domain, which considers the frequency-dependence, as well as the multi-layer soil and the mutual coupling.

$$\begin{cases} \mathbf{Y}_r * \mathbf{V} - \mathbf{A}\mathbf{I} = \mathbf{I}_s \\ \mathbf{Z}_a * \mathbf{I} + \mathbf{A}^T \mathbf{V} = 0 \end{cases} \quad (1)$$

where  $\mathbf{I}_s$  is the current source vector,  $\mathbf{A}$  is the incidence matrix,  $\mathbf{Y}_r$  and  $\mathbf{Z}_a$  are respectively the admittance and the impedance, which are full matrixes. By the vector fitting method [4], each of the frequency-dependent elements can be approximated as a rational function in the complex frequency ( $s$ ) domain:

$$\begin{cases} \mathbf{Y}_r = s\mathbf{C}_r + \mathbf{G}_r + \sum_{q=1}^{Q_r} [\mathbf{k}_r^q / (s + \mathbf{p}_r^q)] \\ \mathbf{Z}_a = s\mathbf{L}_a + \mathbf{R}_a + \sum_{q=1}^{Q_a} [\mathbf{k}_a^q / (s + \mathbf{p}_a^q)] \end{cases} \quad (2)$$

Then, by the inverse Laplace transformation, (1) can be transferred from the frequency domain into the time domain as:

$$\begin{cases} \mathbf{C}_r d\mathbf{v}/dt + \mathbf{G}_r \mathbf{v} + \mathbf{B}_r - \mathbf{A}\mathbf{i} = \mathbf{i}_s \\ \mathbf{L}_a d\mathbf{i}/dt + \mathbf{R}_a \mathbf{i} + \mathbf{B}_a + \mathbf{A}^T \mathbf{v} = 0 \end{cases} \quad (3)$$

where  $\mathbf{B}_r$  and  $\mathbf{B}_a$  are the infinite integral items resulted, and can be calculated by the recursive convolution method [3].

Lastly, by the ADI difference scheme [4], one traditional time step is split into two sub-time steps as (4) and (5). Then, the implicit and explicit difference schemes are respectively applied.

$$\begin{cases} \mathbf{C}_r \frac{\mathbf{v}^{n+1/2} - \mathbf{v}^n}{\Delta t/2} + \mathbf{G}_r \frac{\mathbf{v}^{n+1/2} + \mathbf{v}^n}{2} + \mathbf{B}_r - \mathbf{A}\mathbf{i}^{n+1/2} = \frac{\mathbf{i}_s^{n+1} + 3\mathbf{i}_s^n}{4} \\ \mathbf{A}^T \mathbf{v}^{n+1/2} + \mathbf{L}_a \frac{\mathbf{i}^{n+1/2} - \mathbf{i}^n}{\Delta t/2} + \mathbf{R}_a \frac{\mathbf{i}^{n+1/2} + \mathbf{i}^n}{2} + \mathbf{B}_a = 0 \end{cases} \quad (4)$$

$$\begin{cases} \mathbf{C}_r \frac{\mathbf{v}^{n+1} - \mathbf{v}^{n+1/2}}{\Delta t/2} + \mathbf{G}_r \frac{\mathbf{v}^{n+1} + \mathbf{v}^{n+1/2}}{2} + \mathbf{B}_r - \mathbf{A}\mathbf{i}^{n+1/2} = \frac{3\mathbf{i}_s^{n+1} + \mathbf{i}_s^n}{4} \\ \mathbf{A}^T \mathbf{v}^{n+1/2} + \mathbf{L}_a \frac{\mathbf{i}^{n+1} - \mathbf{i}^{n+1/2}}{\Delta t/2} + \mathbf{R}_a \frac{\mathbf{i}^{n+1} + \mathbf{i}^{n+1/2}}{2} + \mathbf{B}_a = 0 \end{cases} \quad (5)$$

## III. VALIDATION AND CONCLUSION

A typical tower footing device, shown in Fig.1, is tested and simulated by the proposed method. It can be seen that they have quite a good agreement.

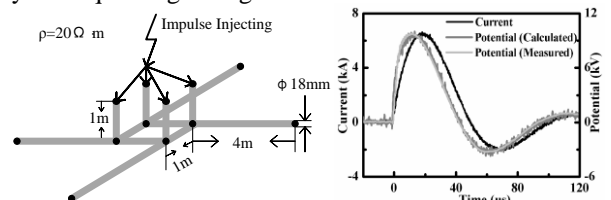


Fig. 1. The validation model and the results

## REFERENCES

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